

# Performance Enhancement of Satellite FSO/QKD Systems using HAP-based Relaying and ARQ

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**Abstract**—This paper proposes to use high-altitude platform (HAP)-based relaying and link-layer automatic repeat request (ARQ) techniques to improve the performance of the satellite quantum key distribution (QKD) over free-space optics (FSO) systems. We also develop analytical frameworks based on two 3-D Markov chain models allowing us to comprehensively analyze the performance of the proposed system in terms of key loss rate and delay outage rate. Our performance analysis takes into account the physical layer impairments induced by FSO channel and receiver noise. Numerical results quantitatively demonstrate that the satellite FSO/QKD system using HAP-based relaying and link-layer ARQ techniques can offer considerable performance improvement over the conventional ones.

**Keywords**—Quantum key distribution (QKD); High-altitude platform (HAP); Free-space optics (FSO); Automatic repeat request (ARQ); Key loss rate (KLR); Delay outage rate.

## I. INTRODUCTION

The desire of connecting everything over Internet requires the secure communication to prevent unauthorized access, stealing, and changing information. Quantum key distribution (QKD), a well-known protocol based on the rule of quantum physics to share the secret key via two lawful parties (namely Alice and Bob) with eavesdropper existed (namely Eve), is a promising method that meets the requirements [1]. A global-scale QKD network can be implemented utilizing satellite-to-ground free-space optics (FSO) links and several proof-of-principle experiments have been performed recently. However, the performance of satellite QKD systems is degraded by atmospheric channels including scattering, absorption, and atmospheric turbulence [2], [3]. Due to these factors, even if there is no eavesdropper existed, quantum key error rate (QKER) may be very high under strong turbulence. To reduce QKER, the reconciliation process based on forward error correction (FEC) has been proposed in [4] and further developed in [5]. Nevertheless, highly computational algorithms are needed for optimizing FEC redundancy and, more importantly, FEC-only techniques are not robust enough to guarantee reliability owing to relatively long-distance transmissions from satellite to ground stations (GSs).

The main objective of this paper is to explore additional techniques to improve performance of satellite FSO/QKD system shown in Fig. 1. We first strengthen the physical layer

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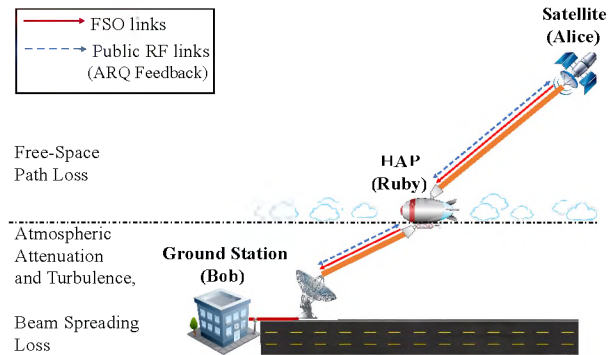


Fig. 1. Satellite FSO/QKD system using HAP-based relaying and link-layer ARQ technique.

by considering a relaying technique based on a high-altitude platform (HAP). Here, a relaying node is located at a HAP that recovers the quantum keys transmitted from a satellite and forwards those keys to a ground station. Thanks to the HAP-based relaying technique, the signal is regenerated and thus the QKER is reduced. Also, we improve the link layer by considering a key-retransmission technique, namely automatic repeat request (ARQ). Unlike FEC technique, which key errors are corrected based on the redundancy added to quantum keys to reduce QKER in the reconciliation process, the link-layer ARQ technique retransmits unsuccessfully keys to assure QKD reliability. ARQ technique does not require high computational algorithms for error control; however, it may cause a high delay because of long transmission distance when keys are retransmitted from the satellite. To solve this problem, we propose to equip a buffer at HAP to store and retransmit the keys when needed. As the distance from HAP to GS is much smaller than the distance from the satellite to GS, the delay is reduced significantly. Finally, crucial performance metrics, including key loss rate and delay outage rate, are analyzed for validating the feasibility of the proposed system. Selected numerical results demonstrate the performance metrics of the proposed system are considerably improved compared to that of the conventional ones.

The rest of this paper is organized as follows. The proposed satellite FSO/QKD system using HAP-based relaying and link-layer ARQ techniques is described in Section II. In Section

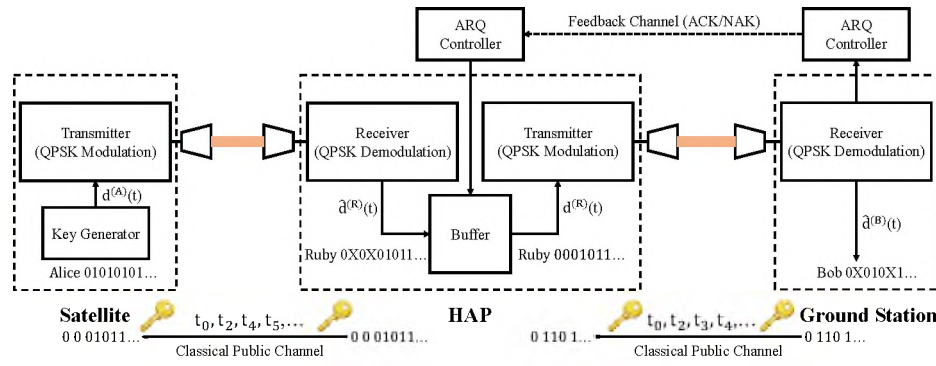


Fig. 2. The proposed satellite FSO/QKD system using HAP-based relaying and link-layer ARQ technique.

III, performance analysis consists of key loss rate and delay outage rate is carried out. In Section IV, selected numerical results are presented. Finally, Section V concludes the paper.

## II. SYSTEM MODEL

### A. QKD Protocol

In this study, we employ continuous variable (CV)-QKD protocol as it is attributed to its capability of providing high key generation rates and especially the compatibility with the standard optical communication technologies. Based on the CV-QKD protocol proposed in [6], Alice encodes the random sequences (i.e., quantum keys) using quadrature phase shift keying (QPSK) signaling and Bob uses dual-threshold (D-T) to detect the signal with the decision rule as

$$\text{Decision} = \begin{cases} 0 & \text{if } (I \geq d_0) \\ 1 & \text{if } (I \leq d_1) \\ X & \text{otherwise} \end{cases} \quad (1)$$

where  $d_0$  and  $d_1$  are two levels of the D-T which symmetrically selected via the “zero” level. Similar to the inaccurate bases choice in conventional BB84 protocol, “X” denotes that Bob does not retrieve Alice’s transmitted bits [7]. Over a radio frequency (RF) public link, time instants when Bob can detect binary bits from received signals, are instantly informed to Alice. Accordingly, Alice also eliminates binary bits corresponds to time instants when Bob detects unidentified bits. Hence, a similar bit sequence is shared between Alice and Bob, is known as *sifted key*. Because of detector/channel imperfections and/or eavesdropping, the sifted key may consist of errors. Similar step as in the original BB84, information reconciliation will be carried out to correct these errors and form error-free *secret key*. However, instead of using FEC, the link-layer ARQ scheme is employed in our considered QKD system.

### B. HAP-based Relaying Technique

In Fig. 2, the proposed satellite FSO/QKD system using HAP-based relaying and link-layer ARQ technique is described, including a satellite (Alice), a HAP-based relaying node (Ruby), and a ground station (Bob) which receives Ruby’s optical signal to recover Alice’s transmitter keys. We

investigate the downlink transmission from Alice’s transmitter placed on the satellite over Ruby’s HAP to Bob’s receiver placed on the ground station. In addition, a classical public channel is used for the link-layer ARQ feedback signal.

In the satellite, the random bit sequence  $d^{(A)}(t)$  created by Alice’s key generator is forwarded to her transmitter. The input bit sequence is modulated at her optical QPSK modulator. The optical signal from Alice’s transmitter is pointed toward Ruby’s optical QPSK demodulator at the HAP. It is significant to note that optical QPSK modulators/demodulators have been described evidently by us in [6]. The transmitted signal is boosted by a telescope that has the gain of  $G_T^{(A)}$  and then sent to Ruby’s HAP via a secure FSO link.

In the HAP-based relaying node, Ruby’s optical signal is firstly passed via the received telescope which has the gain of  $G_R^{(R)}$ . Then, the outputted signal is forwarded to the optical QPSK demodulator to detect the bit sequence. The D-T detector can be used to respectively determine bits “0”, “1”, or “X”. The secret key between Alice and Ruby is retrieved based on the QKD protocol, which is considered in Section II-A.

In the satellite-to-HAP link, the free-space path loss is determined as the main attenuation reduced the received signal quality that leads to QKER can be expected to obtain the low value. Therefore, if Ruby fails to receive the sifted key, then she will discard it immediately without using the FEC technique. When the sifted key is received by Ruby successfully without errors, it will become a secret key that is distributed between Alice and Ruby. Next, the detected key is queued at Ruby’s buffer. Her buffer forwards this key at the front of the queue to her optical QPSK modulator to re-modulate. Currently, the detected key acts as another bit sequence that is passed to Bob. The re-modulated optical signal is also passed via HAP’s transmitted telescope that has the gain of  $G_T^{(R)}$ .

In the ground station, the optical signal is gone via a received telescope that has a gain of  $G_R^{(B)}$  and optical QPSK demodulator. Here, bits are also decided by the D-T detector. In this case, Bob informs both Alice and Ruby about time instants when he can create binary bits. Hence, an identical

bit sequence is shared between Alice, Ruby, and Bob.

### C. Link-layer ARQ Technique

Assuming that atmospheric turbulence mainly occurs in the HAP-to-ground link, to improve our system performance, the link-layer ARQ scheme is deployed between Ruby and Bob. If a sifted key is successfully received by Bob without errors, Bob sends back a local acknowledgment (ACK) to both Ruby and Alice immediately. From Ruby's buffer, she removes the bit sequence which corresponding to the sifted key that has successfully transmitted. If Bob fails to receive the sifted key, Ruby retransmits the bit sequence corresponding to that corrupt sifted key. Denoting  $M$  as the maximum number of retransmission which a bit sequence permitted. If bit sequences are dropped after  $M$  failed attempts, Bob sends a no acknowledgment (NAK) to notify both Ruby and Alice. Consequently, Ruby's buffer eliminates a bit sequence according to cases, either Bob obtains the sifted key with no error (i.e., sifted key is transmitted successfully) or Bob retransmits the bit sequence until  $M$  times, but the respective sifted key is also erroneous.

### III. PERFORMANCE ANALYSIS

Performance indexes are considered including key loss rate (KLR) and delay outage rate. The key loss rate determines the fraction of lost bit sequences in the total number of transmitted ones. Delay outage rate defines the fraction of received bit sequences which delay jitter exceeds the predefined threshold.

#### A. Quantum Key Error Rate

Similar to BB84 protocol, quantum bit error rate (QBER) which is defined as the proportion of erroneous bits of sifted key, can be defined as [8]

$$\text{QBER} = \frac{P_{\text{Error}}}{P_{\text{Sift}}}, \quad (2)$$

where  $P_{\text{Sift}}$  denotes the probability which Ruby (and/or Bob) is able to detect bits (i.e., "0" and "1") based on the D-T detection rule.  $P_{\text{Sift}}$  can be determined as  $P_{\text{Sift}} = P(1,1) + P(0,0) + P(1,0) + P(0,1)$ . However, due to detector/channel imperfections and/or Eve's intervention, Ruby (and/or Bob) may detect unidentified bits (i.e., "X"), thus the probability which Ruby can not decode Alice's transmitted bits, is denoted by  $P_{\text{Error}} = P(1,0) + P(0,1)$ . Where  $P(x,y)$ , with  $x, y \in \{0,1\}$ , represents the joint probability which Ruby (and/or Bob) decodes bit "y" when Alice (and/or Ruby) transmits bit "x", can be expressed as  $P(x,y) = P(x)P(y|x)$ . In which  $P(x) = 1/2$  corresponds to the probability which Alice (and/or Ruby) sends either bit "1" or bit "0".

1) *Satellite-to-HAP Link*: The free-space path loss (FSPL) is determined as the main attenuation reducing the quality of the optical received signal. The joint probabilities of Alice-Ruby via FSO link, is respectively calculated as

$$\begin{cases} P^{(AR)}(x,0) = \frac{1}{2} Q\left(\frac{d_0^{(AR)} - I_x^{(AR)}}{\sigma_N}\right), \\ P^{(AR)}(x,1) = \frac{1}{2} Q\left(\frac{I_x^{(AR)} - d_1^{(AR)}}{\sigma_N}\right), \end{cases} \quad (3)$$

where  $Q(\cdot) \cong \frac{1}{\sqrt{2\pi}} \int_0^\infty \exp(-u^2/2) du$  defines Gaussian Q-function. And  $I_x^{(AR)}$  is the electrical current signals of Ruby (and/or Bob) corresponds to bit "x", is respectively defined as

$$\begin{cases} I_0^{(AR)} = \bar{g} \Re \sqrt{\frac{1}{\text{FSPL}} G_T^{(A)} P_T^{(AR)} G_R^{(R)} P_{LO}}, \\ I_1^{(AR)} = -\bar{g} \Re \sqrt{\frac{1}{\text{FSPL}} G_T^{(A)} P_T^{(AR)} G_R^{(R)} P_{LO}}, \end{cases} \quad (4)$$

where  $P_T^{(AR)}$  is transmitted power.  $P_{LO}$  is the power of local oscillator (LO).  $\bar{g}$ ,  $\Re$  are avalanche multiplication factor, responsivity of the APD, respectively.

Two detection thresholds  $d_0^{(AR)}$  and  $d_1^{(AR)}$  in (3) are determined by using dual-threshold selections as follows

$$\begin{cases} d_0^{(AR)} = E[I_0^{(AR)}] + \varsigma^{(AR)} \sqrt{\sigma_N^2}, \\ d_1^{(AR)} = E[I_1^{(AR)}] - \varsigma^{(AR)} \sqrt{\sigma_N^2}, \end{cases} \quad (5)$$

where  $\varsigma^{(AR)}$  is the D-T scale coefficient.  $E[I_x^{(AR)}]$  is the mean value of  $I_x^{(AR)}$ .  $\sigma_N$  is the total noise variance as follows

$$\sigma_N^2 = 2q_e \bar{g}^{2+x} (\Re P_{LO} + I_d) \Delta f + \frac{4k_B T}{R_L} \Delta f, \quad (6)$$

where  $q_e$ ,  $I_d$ ,  $x$ ,  $T$ ,  $k_B$ , and  $R_L$  denote the electron charge, dark current, excess noise factor, receiver temperature, Boltzmann constant, and load resistance, respectively.  $\Delta f = R_b/2$  denotes the receivers bandwidth with  $R_b$  is system's bit rate.

Next, we determine the quantum key error rate,  $\text{QKER}^{(AR)}$ , where keys are formed with  $l_{bs} P_{\text{Sift}}^{(AR)}$  received bits. The key error probability is expressed as

$$\text{QKER}^{(AR)} = 1 - (1 - \text{QBER}^{(AR)})^{l_{bs} P_{\text{Sift}}^{(AR)}}, \quad (7)$$

where  $l_{bs}$  is the length of a random bit sequence.

2) *HAP-to-Ground Link*: The channel model is composed of three terms including atmospheric attenuation ( $h^l$ ), beam spreading loss ( $h^p$ ), and atmospheric turbulence ( $h^t$ ).

The joint probabilities of Ruby-Bob are expressed as

$$\begin{cases} P^{(RB)}(x,0) = \frac{1}{2} \int_0^\infty Q\left(\frac{d_0^{(RB)} - I_x^{(RB)}}{\sigma_N}\right) f_{h^t}(h^t) dh^t, \\ P^{(RB)}(x,1) = \frac{1}{2} \int_0^\infty Q\left(\frac{I_x^{(RB)} - d_1^{(RB)}}{\sigma_N}\right) f_{h^t}(h^t) dh^t, \end{cases} \quad (8)$$

where

$$\begin{cases} I_0^{(RB)} = \bar{g} \Re \sqrt{G_T^{(R)} P_T^{(RB)} h^l h^p h^t G_R^{(B)} P_{LO}}, \\ I_1^{(RB)} = -\bar{g} \Re \sqrt{G_T^{(R)} P_T^{(RB)} h^l h^p h^t G_R^{(B)} P_{LO}}, \end{cases} \quad (9)$$

where  $P_T^{(RB)}$  is the transmitted power at Ruby's transmitter. In this case, the dual-threshold rule is also used with the D-T scale coefficient  $\varsigma^{(RB)}$ . The quantum key error rate is expressed as

$$\text{QKER}^{(RB)} = 1 - (1 - \text{QBER}^{(RB)})^{l_{bs} P_{\text{Sift}}^{(AR)} P_{\text{Sift}}^{(RB)}}, \quad (10)$$

It is noted that free-space loss, atmospheric attenuation, beam spreading loss, and atmospheric turbulence are considered evidently in [10]. We also use the same set of system parameters, except satellite altitude  $H_S = 600$  km, ground station altitude  $H_G = 5$  m,  $C_n^2 = 5 \times 10^{-15}$  and  $7 \times 10^{-12}$  for weak and strong turbulence, respectively.

TABLE I  
STATE TRANSFORMATION PROBABILITY OF THE QKD  
QUEUE-CONNECTED D-TMC.

Current state	Next state	Transformation probability
(0, B, 0)	(1, B, 0)	$p^{(ar)}p_{BB}$
	(1, G, 0)	$p^{(ar)}p_{BG}$
	(0, B, 0)	$p^{(no)}p_{BB}$
	(0, G, 0)	$p^{(no)}p_{BG}$
$(n, B, m)$ $n \in [1, C-1]$ $m \in [0, M-1]$	$(n+1, B, m+1)$	$p^{(ar)}p_{BB}$
	$(n+1, G, m+1)$	$p^{(ar)}p_{BG}$
	$(n, B, m+1)$	$p^{(no)}p_{BB}$
	$(n, G, m+1)$	$p^{(no)}p_{BG}$
$(n, B, M)$ $n \in [1, C-1]$	$(n, B, 0)$	$p^{(ar)}p_{BB}$
	$(n, G, 0)$	$p^{(ar)}p_{BG}$
	$(n-1, B, 0)$	$p^{(no)}p_{BB}$
	$(n-1, G, 0)$	$p^{(no)}p_{BG}$
$(C, B, m)$ $m \in [0, M-1]$	$(C, B, m+1)$	$p_{BB}$
	$(C, G, m+1)$	$p_{BG}$
$(C, B, M)$	$(C-1, B, 0)$	$p_{BB}$
	$(C-1, G, 0)$	$p_{BG}$
(0, G, 0)	(1, B, 0)	$p^{(ar)}p_{GB}$
	(1, G, 0)	$p^{(ar)}p_{GG}$
	(0, B, 0)	$p^{(no)}p_{GB}$
	(0, G, 0)	$p^{(no)}p_{GG}$
$(n, G, m)$ $n \in [1, C-1]$ $m \in [0, M]$	$(n, B, 0)$	$p^{(ar)}p_{GB}$
	$(n, G, 0)$	$p^{(ar)}p_{GG}$
	$(n-1, B, 0)$	$p^{(no)}p_{GB}$
	$(n-1, G, 0)$	$p^{(no)}p_{GG}$
$(C, G, m)$ $m \in [0, M]$	$(C-1, B, 0)$	$p_{GB}$
	$(C-1, G, 0)$	$p_{GG}$

### B. Link-State Transformation

The two-state Markov Chain is proposed to describe the FSO link-state transformation from Ruby to Bob. The model is widely acknowledged as a mathematically flexible approach for wireless channels [11]. In the proposed Markov model, we discretize the certain time into slots, a slot interval corresponds to the time sent each bit sequence. The link alternates between *bad* (B) and *good* (G) states. A state is considered as good when all sifted keys are transmitted with error-free. If all transmissions are failed, this link-state is bad. The transformation probabilities of link-states are respectively calculated as

$$\begin{cases} p_{GG} = (1 - \text{QKER}^{(AR)})(1 - \text{QKER}^{(RB)})\left(1 - \frac{\tau_{bs}}{\tau_0}\right), \\ p_{BB} = (1 - \text{QKER}^{(AR)})\text{QKER}^{(RB)}\left(1 - \frac{\tau_{bs}}{\tau_0}\right), \end{cases} \quad (11)$$

where  $\tau_{bs} = \frac{l_{bs}}{R_b}$  is the transmission time of a bit sequence.  $\tau_0 = \frac{\sqrt{\lambda L}}{w}$ , is atmospheric turbulence coherent time defined as the time interval that same scintillation coefficient is maintained.  $L$  is the transmission distance in the atmospheric environment.  $w$  is average wind speed,  $\lambda$  is the optical wavelength.

### C. Key Loss Rate

At Ruby's receiver, decoded bit sequence  $\hat{d}^{(R)}(t)$  is forwarded to her buffer with an arrival rate of  $H$  (bit sequences/second). The stationary Bernoulli process is used to model the arrival process of the bit sequence.  $p^{(ar)} = (1 - \text{QBER}^{(AR)})H\tau_{bs}$  and  $p^{(no)} = (1 - \text{QBER}^{(AR)})(1 - H\tau_{bs})$  are the probability when a bit sequence and no bit sequence

arriving in a certain time slot, respectively. In a given time slot, the FSO link keeps stable at its current state. A bit sequence is forwarded to Ruby's transmitter at beginning of a time slot if the queue is not empty. The bit sequence is removed at the ending of each time slot if this transmission is successful.

At the beginning of a time slot, the three-dimensional Discrete-Time Markov Chain (D-TMC) can be determined by  $(n_L, s_L, m_L)$ . In which  $n_L \in [0, C]$  is the number of bit sequences queued at Ruby's buffer, QKD link-state is denoted by  $s_L \in \{B, G\}$ ,  $m_L \in [0, M]$  represents the number of retransmission with a target bit sequence. Our proposed D-TMC is called QKD *queue-connected* D-TMC. Especially, states have both parameters  $m_L > 1$  and  $n_L = 0$  can not obtain. The state transformation of this D-TMC is clearly shown in Tab. I. The probability of the steady state of our QKD *queue-connected* D-TMC, denoted by  $\pi(n_L, s_L, m_L)$ , is determined by solving the balance equation as follows

$$\begin{cases} \Pi^T P_L = \Pi^T, \\ \sum_{n_L=0}^C \sum_{s_L \in \{B, G\}} \sum_{m_L=0}^M \pi(n_L, s_L, m_L) = 1, \end{cases} \quad (12)$$

where  $\Pi = [\pi(n_L, s_L, m_L)]$ .  $P_L$  is the transformation matrix.

The key loss rate caused by failed retransmissions as well as Ruby's buffer overflowed is expressed as follows

$$\text{KLR} = \sum_{n_L=0}^{C-1} \pi(n_L, B, M) + \sum_{s_L \in \{B, G\}} \sum_{m_L=0}^M \pi(C, s_L, m_L). \quad (13)$$

### D. Delay Performance Analysis

Considering time instants when a bit sequence arrives at Ruby's buffer, another three-dimensional Discrete-Time Markov Chain called QKD *bit sequence-connected* D-TMC is developed.  $(n_D, s_D, m_D)$  denotes the bit sequence's current state, with  $n_D \in [1, q_D]$ ,  $s_D \in \{B, G\}$ , and  $m_D \in [0, M]$ . The number of bit sequences that are queued at Ruby's buffer except the currently served one is denoted by  $n_D - 1$ .  $s_D$  and  $m_D$  are the current link-state and the number of retransmissions of a current target bit sequence.  $q_D$  is Ruby's queue length consists of the served bit sequence. At the end of D-TMC process, an absorbing state (i.e., *failure* or *success*), corresponds to the current bit sequence is dropped after  $M$  times failed attempts or retrieved without errors, is determined. Tab. II represents all transformation probabilities of the one-step state.

Denoting  $(q_D, s_L, m_L)$  as the initial state connected with the currently served bit sequence.  $p_{j(q_D, s_L, m_L), (Success)}$  defines the probability which a bit sequence can be successfully received at the  $j$ -th time slot. The conditional probability which queued delay of each obtained bit sequence is greater than  $D/\tau_{bs}$  time slots can be expressed as

$$\Pr\{t_Q > \frac{D}{\tau_{bs}} | (q_D, s_L, m_L)\} = \frac{\sum_{j=D/\tau_{bs}+2}^{q_D(M+1)} p_{j(q_D, s_L, m_L), (Success)}}{\sum_{j=q_D}^{q_D(M+1)} p_{j(q_D, s_L, m_L), (Success)}}. \quad (14)$$

where  $D$  denotes the maximum level of delay jitter which link-layer ARQ requirements allowed.  $t_Q$  describes the queued delay of Ruby's buffer.

TABLE II  
STATE TRANSFORMATION PROBABILITY OF THE QKD BIT  
SEQUENCE-CONNECTED D-TMC.

Current state	Next state	Transformation probability
$(q_D, B, m_L)$	$(q_D, B, m_L + 1)$	$p_{BB}$
$q_D \in (1, B], m_L \in [0, M)$	$(q_D, G, m_L + 1)$	$p_{BG}$
$(q_D, B, M)$	$(q_D - 1, B, 0)$	$p_{BB}$
$q_D \in (1, B]$	$(q_D - 1, G, 0)$	$p_{BG}$
$(1, B, m_L)$	$(1, B, m_L + 1)$	$p_{BB}$
$m_L \in [0, M)$	$(1, G, m_L + 1)$	$p_{BG}$
$(1, B, M)$	Failure	1
$(q_D, G, m_L)$	$(q_D - 1, B, 0)$	$p_{GB}$
$q_D \in (1, B], m_L \in [0, M]$	$(q_D - 1, G, 0)$	$p_{GG}$
$(1, G, m_L)$	Success	1
$m_L \in [0, M]$		

Based on the steady state probability of our proposed QKD queue-connected D-TMC, we can obtain the probability of each initial state connected with a considered bit sequence as follows.  $\Pr\{(q_D, s_L, m_L)\} = \pi(q_D - 1, s_L, m_L)/\pi_v$ , in which  $\pi_v = \sum_{s_L \in \{B, G\}} \pi(0, s_L, 0) + \sum_{q_D=2}^C \sum_{s_L \in \{B, G\}} \sum_{m_L=0}^M \pi(q_D - 1, s_L, m_L)$ . Where  $\pi(q_D - 1, s_L, m_L)$  can be expressed from Section III-C.

Finally, the probability of queued delay exceeding  $D/\tau_{bs}$  time slots can be formulated as

$$\begin{aligned} \Pr\{t_Q > \frac{D}{\tau_{bs}}\} &= \Pr\{t_Q > \frac{D}{\tau_{bs}} | (q_D, s_L, m_L)\} \Pr\{(q_D, s_L, m_L)\} \\ &= \frac{1}{\pi_v} \left[ \sum_{s_L \in \{B, G\}} \Pr\{t_Q > \frac{D}{\tau_{bs}} | (0, s_L, 0)\} \pi(0, s_L, 0) \right. \\ &\quad \left. + \sum_{q_D=2}^C \sum_{s_L \in \{B, G\}} \sum_{m_L=0}^M \Pr\{t_Q > \frac{D}{\tau_{bs}} | (q_D - 1, s_L, m_L)\} \right. \\ &\quad \left. \times \pi(q_D - 1, s_L, m_L) \right]. \end{aligned} \quad (15)$$

#### IV. NUMERICAL RESULTS

This section represents analytical results to determine the proposed FSO/QKD system performance with the selection of parameters listed in Tab. III. We also compare the performance of three systems: (i) using both HAP-based relaying and link-layer ARQ (ARQ/HAP), (ii) using only link-layer ARQ (ARQ), and (iii) without using HAP-based relaying and link-layer ARQ. It would be noted that the later system is the *conventional* one, while the former systems are the proposed ones. Especially, the transmitted power in the systems without using relaying,  $P_T$ , is installed equally to the total required transmitted power from the satellite and HAP in the system using relaying, i.e.,  $P_T = P_T^{(AR)} + P_T^{RB}$ .

##### A. Key Loss Rate

Figure 3 investigates key loss rate (KLR) performance versus arrival rate  $H$  under (a) weak and (b) strong turbulence conditions. Firstly, we confirm that the KLR of the conventional system is relatively high, although the satellite's TX telescope gain and ground station's RX one can be respectively boosted to 130 dB as well as 135 dB, due to impacts of atmospheric turbulence. KLR can be improved

TABLE III  
SYSTEM PARAMETERS

Name	Symbol	Value
<b>General parameters</b>		
Boltzmann constant	$k_B$	$1.38 \times 10^{-23}$ W/K/Hz
Electrons charge	$q_e$	$1.6 \times 10^{-19}$ C
Load resistor	$R_L$	$50 \Omega$
Receiver temperature	$T$	298 K
Excess noise factor	$x$	0.8 (InGaAS APD)
System's bit rate	$R_b$	10 Gbps
Wavelength	$\lambda$	1550 nm
Wind speed	$w$	21 m/s
Avalanche multiplication factor	$\bar{g}$	10
Dark current	$I_d$	3 nA
Responsivity of the APD	$\Re$	0.8
<b>Physic-layer parameters</b>		
Power of local oscillator	$P_{LO}$	0 dBm
Transmitted power	$P_T$	7 dBm
Satellite's TX telescope gain	$G_T^{(A)}$	115 dB
HAP's TX telescope gain	$G_T^{(R)}$	5 dB
HAP's RX telescope gain	$G_R^{(R)}$	105 dB
Ground station's RX telescope gain	$G_R^{(B)}$	5 dB
D-T scale coefficient		
Satellite-to-HAP	$\zeta^{(AR)}$	0.4
HAP-to-ground (Weak)	$\zeta_W^{(RB)}$	0.7
HAP-to-ground (Strong)	$\zeta_S^{(RB)}$	1.4
<b>Link-layer parameters</b>		
Length of bit sequence	$l_{bs}$	$5 \times 10^7$ bit
Maximum delay jitter	$D$	80 ms

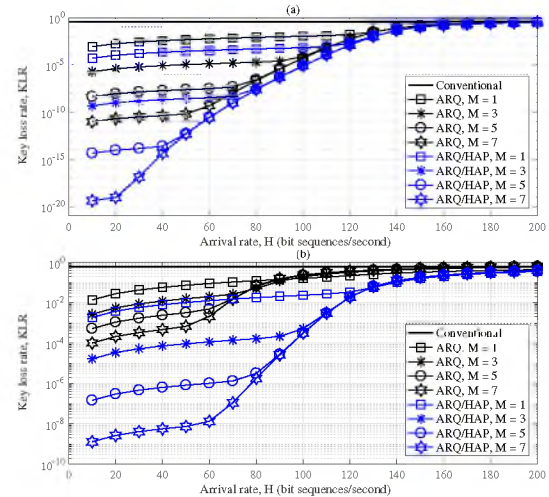


Fig. 3. The key loss rate (KLR) versus arrival rate ( $H$ ) under (a) weak and (b) strong turbulence conditions, when  $C = 10$  (bit sequences).

thanks to the use of link-layer ARQ especially when it is combined with HAP-based relaying. The advantage of using the combined techniques, i.e., both HAP-based relaying and ARQ, compared to ARQ only can be observed evidently, especially for a higher value of  $M$ . In Fig. 3(b) with  $M = 3$ , the maximum arrival rates corresponding to  $KLR = 10^{-2}$  are 120 (sequences/second) for ARQ/HAP case and 50 (sequences/second) for ARQ case. However, using combined techniques can only improve the KLR performance in the



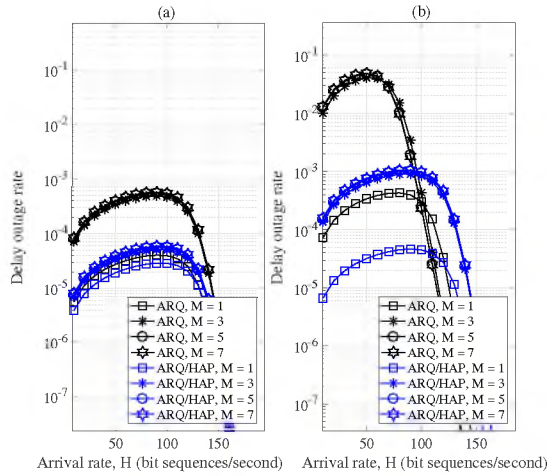


Fig. 4. The delay outage rate versus arrival rate ( $H$ ) under (a) weak and (b) strong turbulence conditions, when  $C = 10$  (sequences).

low arrival rate. When the arrival rate reaches a certain level, the use of combined techniques does not bring the advantage because the probability that Ruby's buffer is full increases. The utilization of combined techniques is recommended when  $H < 180$  for both turbulence conditions.

#### B. Delay Outage Rate

The delay outage rate versus arrival rate  $H$  under different turbulence conditions is considered in Fig. 4. We can realize the delay outage rate in the case of ARQ/HAP is lower than that of ARQ cases, especially in strong turbulence conditions. This is because HAP-based relaying is an efficient method for reducing the key error rate and thus the number of times required for retransmission, which causes the delay. In addition, the increase of  $M$  also causes more delay.

Figure 4 also shows the delay outage rate clearly varies with the arrival rate. In the beginning, because of a higher arrival rate leads to a higher queuing delay, therefore the delay outage increases with the arrival rate. The delay decreases when the arrival rate reaches a certain level. However, in this case, bit sequences are discarded due to buffer overflow, and hence the ratio of successfully received bit sequences decreases.

Figure 5 depicts the delay outage rate versus Ruby's buffer size  $C$  under turbulence conditions. It is clear that longer buffer size helps to reduce KLR; however, it causes more delay. In order to ensure the delay outage rate below a certain threshold, we need significantly to derive the maximum level of Ruby's buffer size. In the case of ARQ/HAP and strong turbulence conditions, the maximum buffer size is 10 sequences with the delay outage rate of  $10^{-3}$ . This value corresponding to different values of  $M$  is trivial under weak turbulence conditions, while they are much different under strong turbulence conditions. The reason is that strong turbulence conditions cause high KLR and thus the number of retransmissions has more impact. With the same buffer size, clearly, the delay outage rate is smaller for the case of weak turbulence conditions.

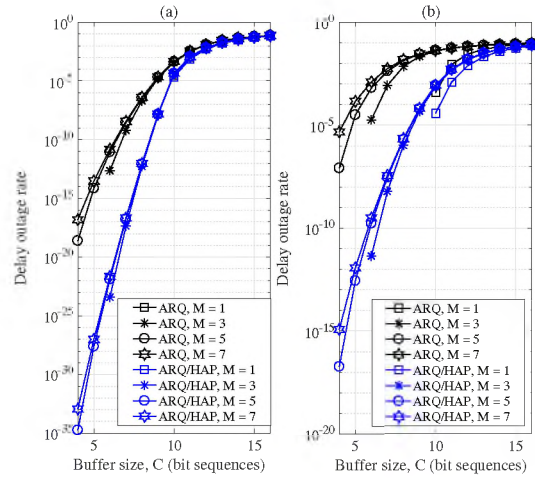


Fig. 5. The delay outage rate versus Ruby's buffer size ( $C$ ) under (a) weak and (b) strong turbulence conditions, when  $H = 60$  (sequence/second).

#### V. CONCLUSION

We investigated the use of HAP-based relaying and link-layer ARQ technique for performance improvement of satellite FSO/QKD systems. We also derived the mathematical expressions for two important performance metrics, the key loss rate and the delay outage rate, of the proposed systems. Numerical results showed that the combination of HAP-based relaying and ARQ helps to significantly improve the system performance, especially under strong atmospheric turbulence conditions. Also, the proper values of the arrival rate and the buffer size can be determined so as to the delay outage rate meets a specific requirement.

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